# マイクロ波測定

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1<sup>st</sup> 2013/09/25 L<sup>st</sup> 2024/02/05

#### Introduction



In microwave engineering, standing wave pattern measurement system, circuit network analyzer, and combination of spectrum analyzer and

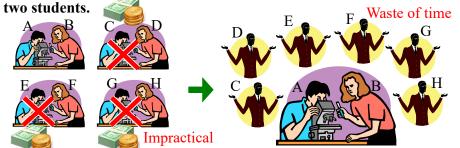
#### Specialized measurement <sup>re used</sup>

system is expensive



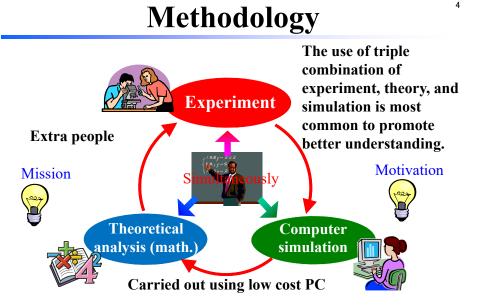
**Motivation / Aims** 

Measurement instruments should be available to equip every one or

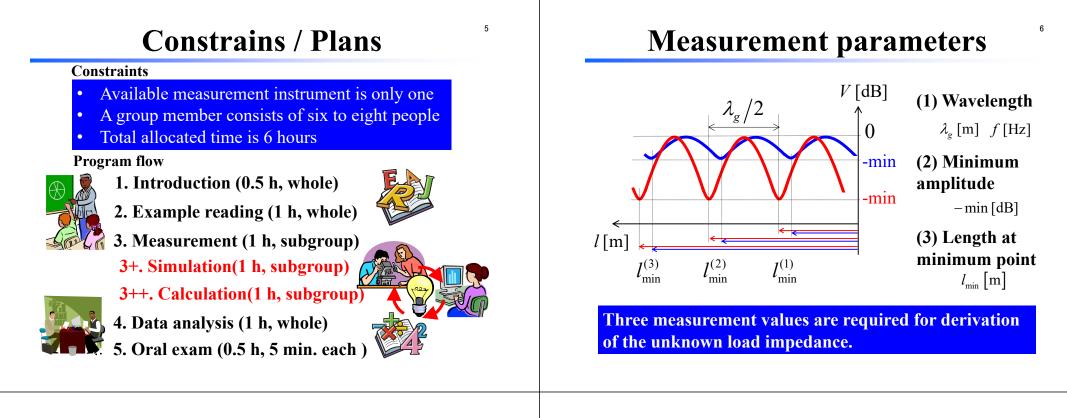


Simultaneously sharing a single precision measurement instrument among a whole group is also unworkable.

# How do we manage a grope within the constraints of time and money ?

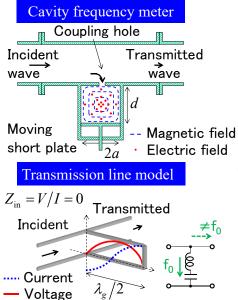


The load on the teaching assistant will increase with increasing number of sub-themes.



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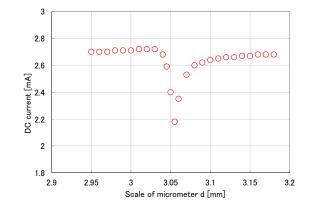
#### **Frequency measurement**



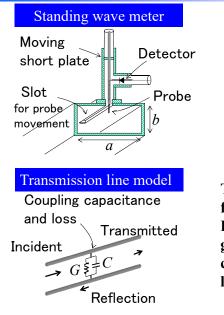


When an incident wave is coupled with a hole in half-wave resonance, the input impedance measured at the incident port becomes zero. When this occurs, the incident wave is completely reflected back to the source.

#### **Measured resonant current**



## **Electric field probe**

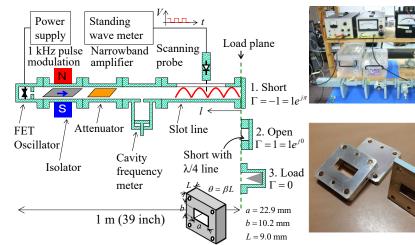




The waveguide center is slotted for electric probe movement. However, the inserted probe will generate a shunt capacitance and conductance in the transmission line.

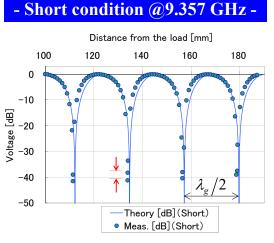
### **Experimental setup (detailed)**

#### - Standing wave pattern measurement system-



User's manual of microwave experimental instrument. 14T150A, SPC Electronics Corp.

## **Measured results**



This result appears to show a good agreement with the theory.



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### **Theoretical pattern**

ショート,オープン,整合の各ケースの「とのの理論値は?

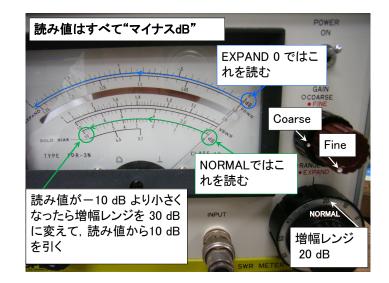
#### 13

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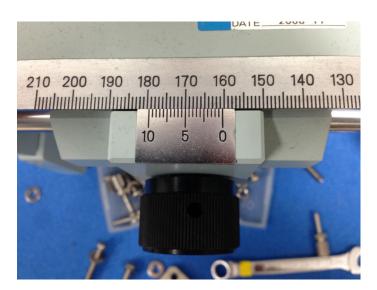
#### **Theoretical parameters**

	Ex. 2.4より抽出	Short (theory)	Open (theroy)	Load (theory)	Ref.
$Z_L[\Omega]$	49.3+j19.7	1.000E-10	1.000E+10	50.0000	
Г	0.0126+j0.1996	-1.0000	1.0000	0.0000	Eq.(6)
<b>Г</b>	0.2000	1.0000	1.0000	0.0000	$=ABS(\Gamma)$
VSWR	1.5000	5.000E+11	2.000E+08	1.0000	Eq.(3)
freq [Hz]	N/A	9.300E+09	9.300E+09	9.300E+09	
$\lambda_0[m]$	N/A	0.0323	0.0323	0.0323	=c/freq
$\lambda_{g}[m]$	0.0400	0.0454	0.0454	0.0454	Eq.(9)
β [rad/m]	157.0796	138.2692	138.2692	138.2692	Eq.(8)
θ[°]	86.4000	180.0000	0.0000	0.0000	
l <sub>min</sub> [m]	0.0148	0.0000	-0.0114	-0.0114	Eq.(5)
$I_{min}/\lambda_{g}$	0.3700	0.0000	-0.2500	-0.2500	

## 定在波増幅器の読み方



定在波測定器の読み方



# 空洞周波数計の読み方



#### **Simulation conditions**

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- OS : Linux (Fedora, CentOS, ...)
- **Software : Gnuplot**
- Wave mode : Plane wave in free space
- **Frequency : 10 GHz**
- Simulation parameter : Γ in complex

$$V(z) = \underbrace{e^{j(\omega t - \beta z)}}_{\text{Incident wave}} + \underbrace{\Gamma e^{j(\omega t + \beta z)}}_{\text{Reflected wave}}$$
(1)

**Reflected wave** Incident wave

$$V(l) = \left| 1 + \left| \Gamma \right| e^{j(\theta - 2\beta l)} \right| = \sqrt{1 + \left| \Gamma \right|^2 + 2\left| \Gamma \right| \cos\left(\theta - 2\beta l\right)}$$
(2)

**Envelope of the standing wave** 

In this step, students are encouraged to create script files.

## **Theoretical analysis**

• Standing wave ratio  

$$SWR = \frac{V_{max}}{V_{min}} = 10^{\frac{|Min[dB]|}{20}} (1)$$
• Absolute value of reflection coefficient  

$$|\Gamma| = \frac{SWR - 1}{SWR + 1} (2)$$
• Phase angle of reflection coefficient  

$$\theta = \pi + 2\beta \left( l_{min} + \frac{n}{2} \lambda_g \right) \qquad n = 0, 1, 2, \cdots (3)$$
• Complex reflection coefficient  

$$\Gamma = |\Gamma| e^{j\theta} \qquad (4)$$
• Normalized load impedance  
• Standing wave ratio  

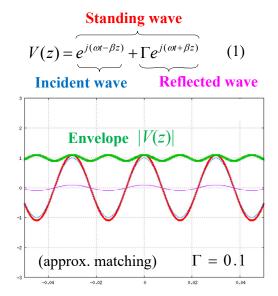
$$F = |\Gamma| e^{j\theta} \qquad (4)$$
• Normalized load impedance

$$z_L = \frac{Z_L}{Z_0} = \frac{1+\Gamma}{1-\Gamma}$$
(5)

normalized impedance using a

scientific electronic calculator.

### Simulated results





Students are instructed to write down the simulated results. Then, they come to understand the image of standing wave.

### **Theoretical analysis**

0.39

#### Excel spread sheet for impedance calculation -

	A	В	C		
1	v 6.3e Jan.2012 Impedance calculation	sheet			
2			Example 2.4	Measured	
3	Light velocity [m/s]	с	3.00E+08		
4	Pai	π	3.141593	values	
5	Frequency [Hz]	f	9.958E+09	Frequency	
6	Waveguide width [m]	а	0.02290	Frequency	
7	Freespace wavelength [m]	λο	0.03013		
8	Guided wavelength [m] (calc.)	λε	0.04000	Wavelength	
10	Guided wavelength [m] (meas.)	λε	0.04000 <	-	
11	Phase constant [rad/m]	ß	157.08	VSWR	
12	Voltage standing wa∨e ratio	SWR	1.50 <		
13	Voltage minimum length [m]	L <sub>min</sub>	0.01480	J I <sub>min</sub>	
	C20 = (1-POWER(C18,2)-POW)	ER(C1	9,2))/(POWI	ER(1-C18,2)	

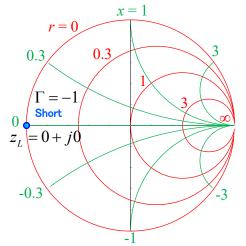
+POWER(C19,2)) 0.95

20	Real part of normalized impedance [ $\Omega$ ]	$Re[z_{L}]$	
21	Imaginary part of normalized impedance [ $\Omega$ ]	Im[zL]	

Impedance is calculated automatically.

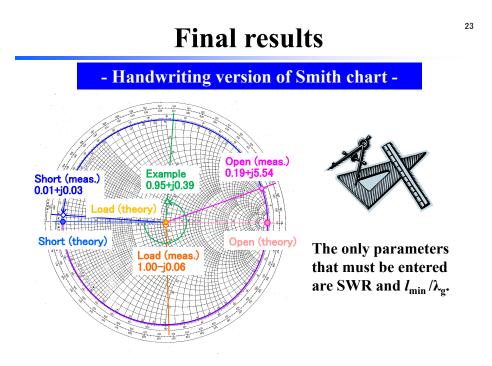
## **Smith Chart**





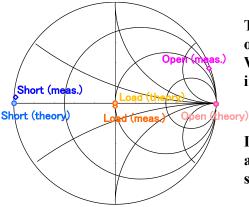
The Smith chart is a polar plot of the reflection coefficient that can be used to derive the load impedance graphically.

Y. Konishi, Y. Sugio & H. Shiomi, (2003). *RF / Microwave circuit CAD and Programing*. Tokyo: K-Laboratory.



### **Analytical results**

#### Plotting on Smith chart by Excel macro-



The Smith chart is a polar plot of the reflection coefficient. We can understand the load impedance schematically.

It seems to be a good agreement is obtained in the short and load conditions.

Y. Konishi, Y. Sugio & H. Shiomi, (2003). *RF / Microwave circuit CAD and Programing*. Tokyo: K-Laboratory.

# 負荷インピーダンスの測定

**Example 2.4 Impedance Measurement with a Slotted Line** The following two-step procedure has been carried out with a 50  $\Omega$  coaxial slotted line to determine an unknown impedance:

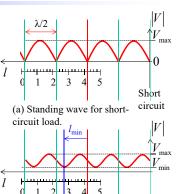
- 1. A short circuit is placed at the load plane, resulting in a standing wave on the line with infinite SWR, and sharply defined voltage minima, as shown in Figure 2.14a. On the arbitrarily positioned scale on the slotted line, voltage minima are recorded at z = 0.2 cm, 2.2 cm, 4.2 cm.
- 2. The short circuit is removed, and replaced with the unknown load. The standing wave ratio is measured as SWR = 1.5, and voltage minima, which are not as sharply defined as those in step 1, are recorded at z = 0.72 cm, 2.72 cm, 4.72 cm, as shown in Figure 2.14b.

Find the load impedance.

#### Solution

Knowing that voltage minima repeat every  $\lambda/2$ , we have from the data of step 1 above that  $\lambda = 4.0$  cm. In addition, because the reflection coefficient and input impedance also repeat every  $\lambda/2$ , we can consider the load terminals to be effectively located at any of the voltage minima locations listed in step 1. Thus, if we say the load is at 4.2 cm, then the data from step 2 shows that the next voltage minimum away from the load occurs at 2.72 cm, giving  $l_{min} = 4.2 \cdot 2.72 = 1.48$  cm = 0.37 $\lambda$ .

D. M. Pozar, Microwave Engineering, 3rd ed., pp.71-72 より引用



(b) Standing wave for unknown load.

$$\begin{split} & \left| \Gamma \right| = 0.2, \, \theta = 86.4^{\circ}, \\ & \Gamma = 0.0126 + j0.1996, \\ & Z_L = 47.3 + j19.7 \, \Omega \end{split}$$

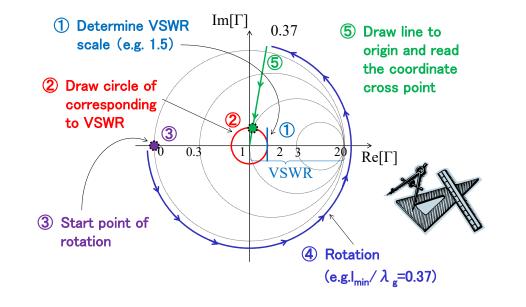
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Unknown

load

スミスチャートプロット手順

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# スミスチャートプロット手順

#### **Example 2.4 Impedance Measurement with a Slotted Line**

For the Smith chart version of the solution, we begin by drawing the SWR circle for SWR=1.5, as shown in Figure 2.15; the unknown normalized load impedance must lie on this circle. The reference that we have is that the load is 0.37 $\lambda$  away from the first voltage minimum. On the Smith chart, the position of a voltage minimum corresponds to the minimum impedance point (minimum voltage, maximum current), which is the horizontal axis (zero reactance) to the left of the origin. Thus, we begin at the voltage minimum point and move 0.37 $\lambda$  toward the load (counterclockwise), to the normalized load impedance point,  $z_L = 0.95 + j0.4$ , as shown in Figure 2.15. The actual load impedance is then  $Z_L = 47.5 + j20\Omega$ , in close agreement with the above result using the equations. Note that, in principle, voltage maxima locations could be used as well as voltage minima positions, but voltage minima are more sharply defined than voltage maxima, and so usually result in greater accuracy.

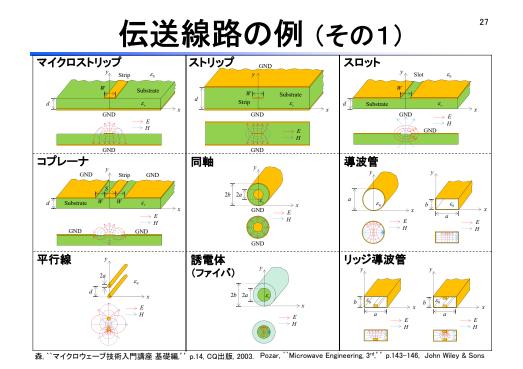
#### VSWRに相当する円の半径を決める。VSWRの値は実軸に目盛られた抵抗分 R/Z0の数値に等しい(例題では1.5)。VSWR > 30のときはほぼ全反射と考えてよ い。当然, VSWR = ∞ のときは完全な全反射である。

② 原点を中心としてVSWRに相当する円を描く。

#### ③ 左端の最小点を回転の始点とする。

④ 始点から波長換算した距離lmin/λg(例題では0.37)だけ負荷方向へ回転させる。
 ⑤ 回転させた位置から原点に向かって直線を引き,交点座標を読む。(例題では zL=0.95+ j 0.4が得られればOK)

D. M. Pozar, Microwave Engineering, 3<sup>rd</sup> ed., pp.72-72 より引用



#### **Oral exams**

It is important to confirm the final level of understanding of each student. To this end, the following oral questions are given to each student:



#### • What is a <u>standing wave</u>?

- What is load impedance? What does <u>short, open or</u> load mean in reference to the electric circuit?
- <u>What can it do by Smith chart? Which is the reactance circle</u>? Which is the <u>resistance circle</u>? <u>Where is the short, open, load on the Smith chart?</u>
- What is a waveguide? Describe the difference with <u>other transmission lines</u> using an example.